# Performance of circular patch microstrip antenna for adaptive modulation and coding applications

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## **ABSTRACT**

The thickness of the substrate is one factor influencing antenna performance. In this paper, the circular patch antenna was designed using rogers RT5880 with dielectric constant of 2.20, loss kept 0.0009 and FR-epoxy with dielectric constant of 4.4, loss kept 0.02. The thickness of the substrate was varied to see how it affected antenna performance, such as return loss. The thickness of the checked substrate is 1.58 mm, which is the standard thickness for rogers RT5880 as well as 2.08 mm, 2.58 mm and 2.85 mm. The simulation work is carried out by ANSYS HFSS software. In this paper, the thicknesses of different substrates are checked while other parameters stay constant and the circular patch radius of the antennas is optimized to achieve a resonant frequency of 3.5 GHz based on the thickness of the above-mentioned substrates used. Return loss, VSWR, gain and half power beam width were found, and the results showed that with the increase in the thickness of the substrate, the gain increases and the value of half power beam width (HPBW) and better results were obtained in the case of RTduroid 5880 and for thicknesses h=2.85 mm.

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#### 1. INTRODUCTION

Antennas with a small footprint and the ability to operate across multiple frequency bands have generated a lot of interest in recent years, especially for novel wireless communication systems such as radio frequency identification [1], microwave-based global interoperability and wireless local area networks (WLAN). (WiMAX) [2]–[7], among which are applications modification and adaptive coding. Multiband antennas should operate in the appropriate frequency ranges for these applications: 2.45 GHz (2.40-2.50 GHz), 3.5 GHz (3.40-3.70 GHz), and 5.8 GHz (5.725-5.875 Hz) [8]. The 3.5 GHz frequency was selected for this study. Microstrip antennas offer a wide range of uses, including medical, military, mobile, and satellite communication [9]–[13]. As wireless applications need greater capacity, wideband antennas operating at higher frequencies are unavoidable [14]–[21]. The efficiency of microstrip antennas is low, and their bandwidth is small, and substrate parameters like dielectric constant, homogeneity, and tangent loss impact their performance [22]–[25]. The need for wideband antennas operating at higher frequencies is growing rapidly as more wireless applications require greater bandwidth [26]. Broadside radiators make up the vast bulk of microstrip antennas [27]. Patches of printed antennas' radiating elements come in a diverse range of shapes, including rectangle, square, elliptical, and circular shape [28]–[31]. A suitable model selection can also be used to select an end-fire radiator [32], [33]. At microwave frequencies (f > 1 GHz), a microstrip patch

antenna is one of the most useful antennas. It comprises a metal "patch" above the insulating substrate and a ground plane beneath the insulating layer. The feed position must be altered to adjust the input impedance as previously [34]-[36]. Thicker substrates with lower dielectric constants are more efficient, have a bigger bandwidth, enhance the patch's fringing field and hence radiating power, but add weight and surface wave losses [37]. A patch antenna can be fed in various ways, including probe feed, microstrip line feed, aperture coupled feed, and proximity coupling feed [38]. Saidulu et al. [39] focused on the circular patch-fed axial probe of microstrip antenna properties with and without a dielectric substrate. As the return loss and VSWR increase, the bandwidth and gain decrease with the dielectric constant of the super-substrates. According to Hamzah et al. [40] the circular patch antenna is built using a rogers RT5880 with a tangent loss of 0.0009 and 2.2 is the dielectric constant. The purpose of this study was to see how the thickness of the substrate affected the performance of the antenna, i.e. thickness of that substrate has been modified, resulting in a change in return loss. CST microwave studio conducted the simulations. Research by Sharma et al. [41] used the ultra-high frequency (UHF) band from 470MHz to 806MHz and developed the circular microstrip patch antenna used for this band. The pin from the center is used to shorten the main patch. Sateaa et al. [42] presented (3) rectangular microstrip patch array antennas ((1\*1), (1\*2), and (1\*4)) with easy and small construction. The suggested antenna was designed utilizing (CST) and printed on a rogers RT5880 (lossy) substrate with a dielectric constant of 2.2, a loss tangent of 0.0009, and a thickness of 0.1 mm. The patch size is (1.570 mm\*2 mm) for three designs of microstrip patch array antenna, and it operates in dual-band. Thakur and Singh [43] proposed a novel photonic band gap substrate-based microstrip patch antenna for cancer detection. With a gain of 3.430 dBi and a reflection coefficient of -68.530 dB, the antenna resonates at 0.198 THz. According to Jairath et al. [44] the antenna designed in this study exhibits an impedance bandwidth across the ultrafast bandwidth (UWB) frequency range of 3.10 to 14 GHz with a constant voltage wave ratio of less than 2, excluding the band stops at 3.29 to 3.70GHz (WiMAX band), 3.70 to 4.10 GHz (C band), 5.10 to 5.90GHz (WLAN band), and 7.060 to 7.76 GHz (downlink X band satellite communication). The suggested antenna, which is built on a low-cost FR-4 substrate, has a small footprint of (24\*20\*1.6) mm<sup>3</sup>. Jairath and Singh [45] presented a novel WLAN and X-band frequency rejected antenna with an elliptical slot in the radiating patch and a metamaterial inspired strict source and record route SSRR strip. The antenna reported in this study has a low voltage standing wave ratio at the rejected frequency ranges (5.150-5.825GHz), (7.360-8.06GHz), and (9.54-10.38GHz).

The used high frequency simulation software high-frequency structure simulator (HFSS) program to develop the design, which was set to 3.5 GHz. In the circular antenna, two substrates were employed, with thicknesses of 1.58 mm, 2.08 mm, 2.58 mm and 2.85 mm for each substrate. This paper is organized as: section 2 deals with the proposed circular antenna. The third section shows the method for calculating the parameters of the designed circular antenna. The results are presented and discussed for the designed antenna, for two types of RT/duroid5880 and Fr epoxy substrates with thicknesses of 1.58 mm, 2.05 mm, 2.58 mm and 2.85 mm in the fourth section. The fifth section contains the conclusions.

#### 2. RESEARCH METHOD

The suggested antenna dimensions are computed using the operating frequency, fr, of 3.5 GHz, the height of the substrate of 1.58 mm, 2.05 mm, 2.58 mm and 2.85 mm, the dielectric permittivity of the Fr-epoxy and RT5880 substrates, of 2.20 for RT5880 and 4.4 for Fr-epoxy, as shown in Table 1.

Table 1. Antenna design specifications for microstrip circular patch

Parameters for antenna design	Specifications									
Frequency of operation	3.5 GHz	3.5 GHz	3.5 GHz	3.5 GHz	3.5 GHz	3.5 GHz	3.5 GHz	3.5 GHz		
Substrate	Fr-epoxy (Er=4.4)	Fr-epoxy (&r=4.4)	Fr-epoxy (Er=4.4)	Fr-epoxy (Er=4.4)	RT-duroid 5880. (Er=2.20)	RT-duroid 5880. (Er=2.20)	RT- duroid 5880. (Er=2.20)	RT-duroid 5880. (&=2.20)		
Thickness of substrate(h)	1.58 mm	2.05 mm	2.58 mm	2.85 mm	1.58 mm	2.05 mm	2.58 mm	2.85 mm		
Tangent factor	0.02	0.02	0.02	0.02	0.0009	0.0009	0.0009	0.0009		

Microstrip circular patch antenna using fr epoxy and RT-duroid 5880 is designed using HFSS software and analysis has been carried out to determine the design parameter to achieve the adaptive modulation and coding applications requirements. Figure 1 shows the proposed microstrip circular patch antenna designed using HFSS. Circular geometries have benefits over conventional geometries in some applications, such as arrays, because the feed may be linked at any point along the circumference. The basic antenna design equations are given step by step as:

a. The radius of the circular patch antenna (R)

$$R = \frac{F}{\sqrt{1 + \frac{2h}{\pi \cdot \varepsilon_T \cdot F\left[\ln\left(\frac{F\pi}{2h}\right) + 1.7726\right]}}} \tag{1}$$

$$F = \frac{8.791*10^9}{f_{res}\sqrt{\varepsilon_r}} \tag{2}$$

Where h is substrate height and  $\varepsilon_r$  is substrate dielectric constant

b. Microstrip line width

$$\frac{w}{h} = \begin{cases} \frac{3e^{K}}{e^{2K} - 2} for \frac{w}{h} < 2\\ \frac{2}{\pi} \left[ P - 1 - (2P - 1) + \frac{\varepsilon_{r} - 1}{2\varepsilon_{r}} \left\{ ln(P - 1) + 0.39 - \frac{0.61}{\varepsilon_{r}} \right\} \right] for \frac{w}{h} < 2 \end{cases}$$
(3)

$$K = \frac{z_0}{60} \sqrt{\frac{\varepsilon_r + 1}{2}} + \frac{\varepsilon_r - 1}{\varepsilon_r + 1} \left( 0.23 + \frac{0.11}{\varepsilon_r} \right) \text{ and } P = \frac{377\pi}{2Z_0 \sqrt{\varepsilon_r}}$$
 (4)

where Z<sub>o</sub> is line impedance

Using the in (3) and (4), the width of the 50 line (W<sub>1</sub>) and the width of the transformer line (W<sub>2</sub>) may be determined.

c. Quarter wave transformer length

$$L = \frac{\lambda_d}{4} = \frac{\lambda_o}{4\sqrt{\varepsilon_{ref}}} \tag{5}$$

$$\frac{\varepsilon_{r+1}}{2} + \frac{\varepsilon_{r+1}}{2} \left( 1 + \frac{12h}{w} \right)^{-0.5} \tag{6}$$

Where λo is wavelength of free space and λd is dielectric substrate of wavelength

d. Radiation losses conductance (G<sub>R</sub>)

$$G_R = \frac{2.39}{4.h. f_o.\mu_o.Q_R} \tag{7}$$

$$G_{R} = \frac{2.39}{4.h. f_{o}.\mu_{o}.Q_{R}}$$

$$Q_{R} = \frac{4.R(\alpha_{11}^{2} - 1).\epsilon^{\frac{3}{2}}}{h.\alpha_{11}^{3}.F(\frac{\alpha_{11}}{\sqrt{\varepsilon_{F}}})}$$
(8)

Where  $\varepsilon_r$  is dielectric constant,  $\alpha 11=1.84118$ , and Ris radius of circular patch

e. Dielectric losses conductance (G<sub>D</sub>)

$$G_D = \frac{2.39 \cdot \tan \delta}{4 \mu_0 h_0 f_0} \tag{9}$$

f. Conduction (ohmic) losses conductance (Gc)

$$G_c = \frac{2.39 \,\pi \,. (\pi.\mu_{0.} f_0)}{4 \,h^2 \,\sqrt{\sigma}} \tag{10}$$

Where  $\sigma$  is conductance of copper material used in design (5.8\*10<sup>7</sup>) and  $\mu_{o}=1.256*10^{-6}$ 

$$G_T = G_R + G_D + G_C \tag{11}$$

g. The impedance of quarter wave transformer (Z<sub>tr</sub>)

$$Z_{tr} = \sqrt{Z_T \cdot Z_o} \tag{12}$$

Where Zo is characteristic impedance of feed line and  $Z_{tr}$  is impedance of  $\mathcal{N}4$  transformer

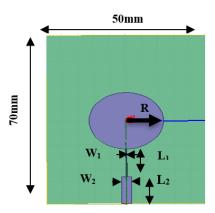


Figure 1. Circular microstrip antenna edge feeding

To match a 50 feed transmission line with the circular microstrip patch, a quarter-wave transformer is used. A quarter-wave transformer is a simple impedance transformer used in impedance matching to decrease the energy reflected when a transmission line is connected to a load. The needed calculations for the circular antenna sections are shown in Table 2.

Table 2. Calculations of circular microstrip antenna with edge feeding for eight substrates

1 41		ulations of Ci	iculai illicio	with edge feeding for eight substrates				
Parameter	Fr epoxy (Er=4.4, h=1.58m m)	Fr epoxy (&r=4.4, h=2.08mm)	Fr epoxy (&r=4.4, h=2.58mm)	Fr epoxy (&r=4.4, h=2.85mm)	RT- duroid 5880. (h=1.58mm , &r=2.20)	RT- duroid 5880. (h=2.08mm , &r=2.20)	RT- duroid 5880. (h=2.58mm , &r=2.20)	RT- duroid 5880. (Er=2.20, h=2.85mm)
Patch radius (mm)	11.7	11.55	11.38	11.25	16.21	16.1	16.05	15.7
Length of the $\lambda/4$ transformer	11.74	12.065	12.109	12.717	16.41	15.52	15.523	16.3012
$(L_1)$ mm Width of the $\lambda/4$	0.253	0.31	0.4236	0.452	0.611	0.877	1.255	1.104
transformer $W_1$ ) mm $50 \Omega$ line length( $L_2$ )	11.183	11.183	11.183	11.1832	14.919	14.919	14.919	14.919
mm $50 \Omega$ line width(W <sub>2</sub> )	3.036	3.976	4.932	5.4532	4.883	6.4088	7.949	8.627
mm GR(Siemen s)	8.853*10-	8.446*10-4	8.446*10-4	8.446*10-4	2.478*10-3	3.377*10-3	3.376*10-3	3.375*10-3
GD (Siemens)	1.7118*1 0 <sup>-3</sup>	1.3069*10 <sup>-3</sup>	1.053*10-3	9.708*10-4	7.7031*10 <sup>-5</sup>	5.881*10-5	4.7413*10 <sup>-5</sup>	4.368* 10-5
GC (Siemens)	6.2413*1 0 <sup>-5</sup>	3.51*10 <sup>-5</sup>	2.281*10-5	1.937*10 <sup>-5</sup>	6.022*10 <sup>-5</sup>	3.51*10 <sup>-5</sup>	2.281*10 <sup>-5</sup>	1.937*10 <sup>-5</sup>
GT (Siemens)	26.5957* 10 <sup>-4</sup>	21.866*10-4	19.2041*10	18.3477*10	2.616*10 <sup>-3</sup>	3.47091*10	3.446*10 <sup>-3</sup>	3.438*10-3
$ZT(\Omega)$	376	457.331	520.722	545.027	382.26	288.108	290.17	290.862
$Ztr(\Omega)$	137.113	151.216	161.357	165.079	138.242	120.022	120.45	120.594

# 3. RESULTS AND DISCUSSION USING HIGH-FREQUENCY STRUCTURE SIMULATOR 3.1. Result of circular microstrip antenna using Fr epoxy (h=1.58 mm, h=2.08, h=2.58mm and h=2.85mm with $\varepsilon_r$ =4.4)

The Figure 2 shows return loss plot for circular microstrip antenna with four thicknesses (h=1.58 mm, h=2.08, h=2.58 mm and h=2.85 mm,  $\varepsilon_r$ =4.4), Where the better return loss value Fr epoxy (h=2.85 mm) is (--22.4776 dB). The VSWR of circular microstrip antenna with four thicknesses are seen in Figure 3. As can be seen in Figure 4 (in appendix) the gain value of circular antenna with four thicknesses, where Figure 4(a) for h=1.58 mm, Figure 4(b) for h=2.05 mm, Figure 4(c) for h=2.58 mm and Figure 4(d) for h=2.85 mm). Figure 5 (in appendix) depicts the circular antenna's radiation pattern, which was used to determine the half-

power beam width, where Figure 5(a) for h=1.58 mm, Figure 5(b) for h=2.08 mm, Figure 5(c) for h=2.58 mm, and Figure 5(d) for h=2.85 mm.

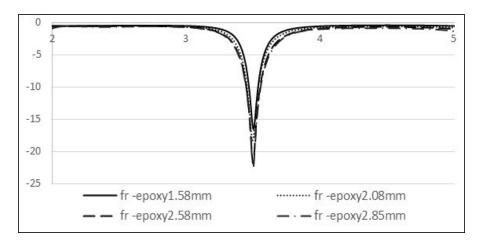


Figure 2. Circular microstrip antenna return loss for Fr-epoxy with €r=4.4

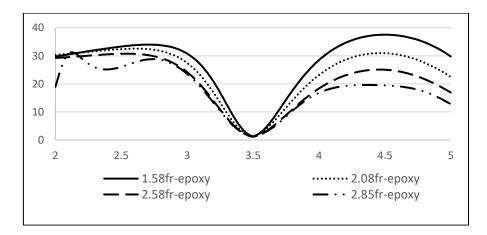


Figure 3. VSWR of circular microstrip antenna for Fr-epoxy with Cr=4.4

# 3.2. Result of circular microstrip antenna using Rogers RT/Duroid5880 (h=1.58 mm, h=2.08 mm, h=2.58 mm and h=2.85 mm with $\varepsilon$ r=2.20)

Figure 6 illustrates a return loss plot for a circular microstrip antenna with four thicknesses (h=1.58 mm, h=2.08, h=2.58 mm and h=2.85 mm, with C=2.20). Figure 7 shows the VSWR of a circular microstrip antenna with four thicknesses (h=1.58 mm, h=2.08, h=2.58 mm and h=2.85 mm, with C=2.20). Figure 8 (in appendix) shows a circular microstrip antenna's gain when fed from the edge for RT Duroid5880 with four thicknesses, where Figure 8(a) for h=1.58 mm, Figure 8(b) for h=2.05 mm, Figure 8(c) for h=2.58 mm and Figure 8(d) for h=2.85 mm). Figure 9 (in appendix) demonstrates a circular microstrip antenna's radiation pattern for RT Duroid5880, Figure 9(a) shows (RT Duroid5880 (h=1.58 mm, C=2.20)), Figure 9(b) shows (RT Duroid5880 (h=2.08 mm, C=2.20)), Figure 9(c) shows (RT Duroid5880 (h=2.58 mm, C=2.20)), and Figure 9(d) shows (RT Duroid5880 (h=2.85 mm, C=2.20)).

Table 3 summarizes the modeling findings for these four substrates at 3.5GHz (in appendix). The circular antenna with substrate rogers RT duroid5880 and the 4 suggested thicknesses has superior performance than the circular antenna with Fr epoxy substrate, as shown in Table 3. It's worth noting that as the substrates' thickness gets thicker, the gain and HPBW improve. The greatest gain is 8.0998 dB for rogers RT duroid5880 (&r=2.2, h=2.85 mm). When the thickness is increased, the return losses and VSWR of the Fr epoxy substrate are better than that of rogers RT duroid5880 substrate. Figures 10 to 17 show the priority of RT duroid5880 over Fr-Epoxy in terms of VSWR, Returnloss for all cases used in the design.

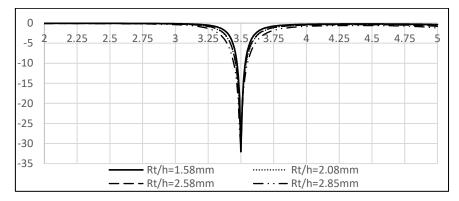


Figure 6. Circular microstrip antenna return loss for RT/Duroid5880 with Cr=2.20

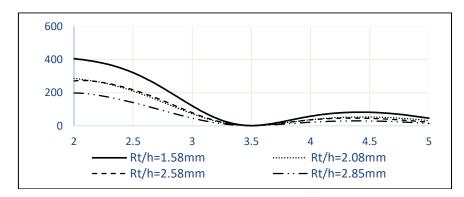


Figure 7. VSWR of circular microstrip antenna for RT duroid5880 with €r=2.20

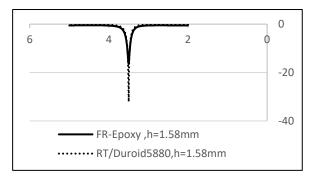


Figure 10. Comparison circular microstrip antenna return loss between Fr-epoxy and RT/duroid5880, (h=1.58 mm)

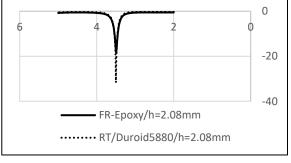


Figure 11. Comparison circular microstrip antenna return loss between Fr-epoxy and RT/duroid5880, (h=2.08 mm)

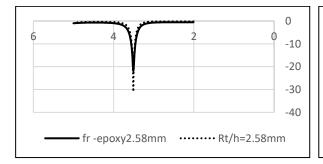


Figure 12. Comparison circular microstrip antenna return loss between Fr-epoxy and RT/duroid5880, (h=2.58 mm)

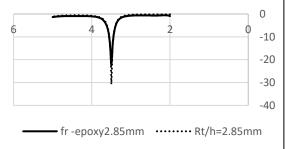
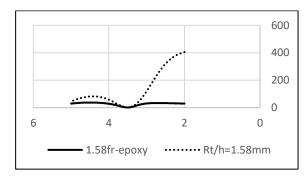


Figure 13. Comparison circular microstrip antenna return loss between Fr-epoxy and RT/duroid5880, (h=2.85 mm)



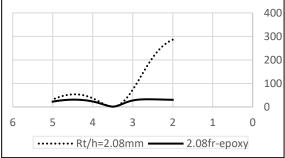
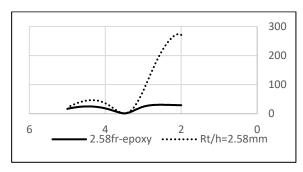


Figure 14. Comparison of VSWR for circular microstrip antenna between Fr-epoxy and RT/duroid5880, (h=1.58 mm)

Figure 15. Comparison of VSWR for circular microstrip antenna between Fr-epoxy and RT/duroid5880, (h=2.08 mm)



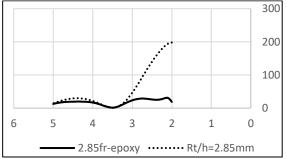


Figure 16. Comparison of VSWR for circular microstrip antenna between Fr-epoxy and RT/duroid5880, (h=2.58 mm)

Figure 17. Comparison of VSWR for circular microstrip antenna between Fr-epoxy and RT/duroid5880, (h=2.85 mm)

#### 4. CONCLUSION

The HFSS 15.0 simulation program is used to create a circular antenna that resonates at 3.5GHz. Eight circular microstrip antennas were designed with Fr epoxy (Er=4.4, h=1.58 mm), Fr epoxy (Er=4.4, h=2.08 mm), Fr epoxy (Er=4.4, h=2.58 mm), Fr epoxy (Er=4.4, h=2.85 mm), Rogers RT/duroid5880 (Er=2.20, h=1.58 mm), Rogers RT/duroid5880 (Er=2.20, h=2.58 mm), Rogers RT/duroid5880 (Er=2.20, h=2.58 mm) and Rogers RT/duroid5880 antennas (Er=2.20, h=2.85 mm). The results indicated the superior performance of the circular antenna with Rogers RT/duroid5880 (Er=2.20., h=2.85 mm) in terms of gain and HPBW, the gain value equals 8.0998db and HPBW equals 63.2888dB.

#### **ACKNOWLEDGEMENTS**

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## **APPENDIX**

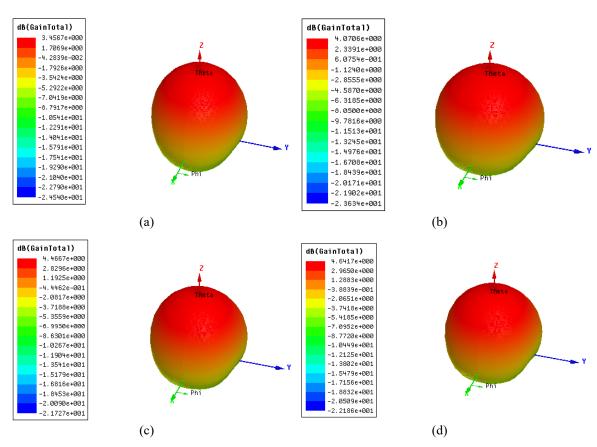


Figure 4. Gain of circular microstrip antenna (a) (Fr epoxy (h=1.58 mm,  $\varepsilon$ r=4.4)), (b) (Fr epoxy (h=2.08 mm,  $\varepsilon$ r=4.4)), (c) (Fr epoxy (h=2.58 mm,  $\varepsilon$ r=4.4)) and (d) (Fr epoxy (h=2.85 mm,  $\varepsilon$ r=4.4))

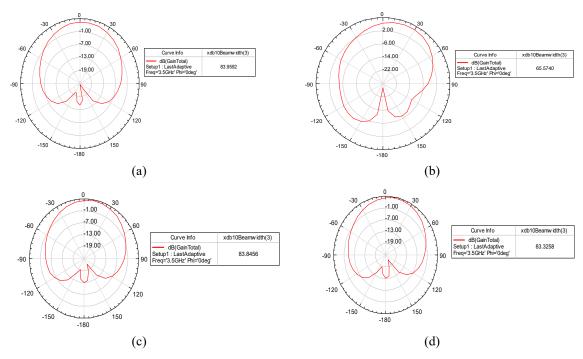


Figure 5. Radiation pattern of circular microstrip antenna, (a) (Fr epoxy (h=1.58 mm, Cr=4.4)), (b) (Fr epoxy (h=2.08 mm, Cr=4.4)), (c) (Fr epoxy (h=2.58 mm, Cr=4.4)) and (d) (Fr epoxy (h=2.85 mm, Cr=4.4))

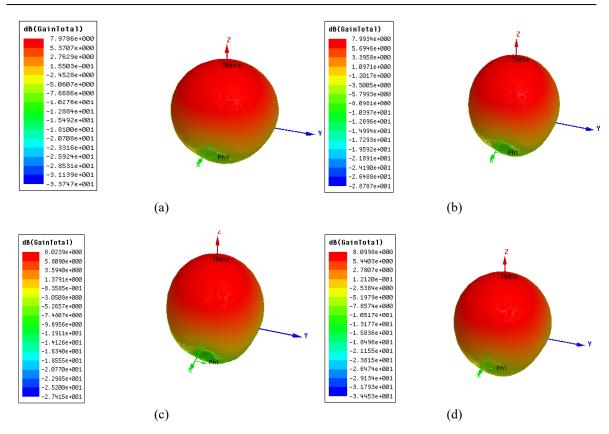


Figure 8. Gain of circular microstrip antenna (a) RT duroid5880 (h=1.58 mm,  $\varepsilon$ r=2.20), (b) RT duroid5880 (h=2.08 mm,  $\varepsilon$ r=2.20), (c) RT duroid5880 (h=2.58 mm,  $\varepsilon$ r=2.20), and (d) RT duroid5880 (h=2.85 mm,  $\varepsilon$ r=2.20)

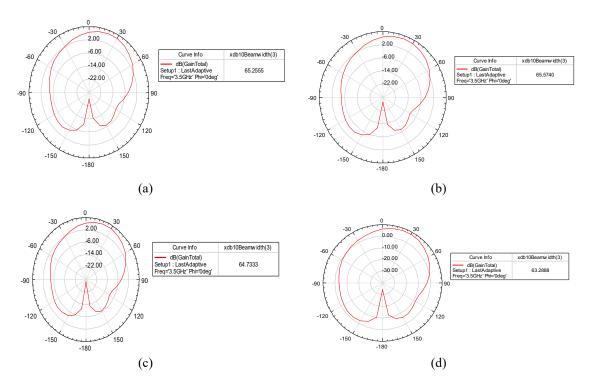


Figure 9. Radiation pattern of circular microstrip antenna, (a) (RT duroid5880 (h=1.58 mm,  $\varepsilon$ r=2.2)), (b) (RT duroid5880 (h=2.08 mm,  $\varepsilon$ r=2.20)), (c) (RT Duroid5880 (h=2.58 mm,  $\varepsilon$ r=2.20)) and (d) (RT duroid5880 (h=2.85 mm,  $\varepsilon$ r=2.20))

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Parameter	Fr epoxy (Er=4.4, h=1.58 mm)	Fr epoxy (&=4.4, h=2.08 mm)	Fr epoxy (&=4.4, h=2.58 mm)	Fr epoxy (Er=4.4, h=2.85 mm)	RT- duroid 5880. (h=1.58 mm, Er=2.20)	RT- duroid 5880. (h=2.08 mm, & Er=2.20)	RT- duroid 5880. (h=2.58 mm ,£r=2.20)	RT- duroid 5880. (Er=2.2, h=2.85 mm)
VSWR	1.3518	1.2848	1.1734	1.1626	1.0513	1.0535	1.0585	1.0599
Return loss (dB)	-16.5015	-18.0866	-21.2606	-22.4776	-32.0420	-31.6891	-30.9252	-30.7227
Gain (dB)	3.4567	4.0706	4.4667	4.6417	7.9786	7.9934	8.0239	8.0998
Half power beam width(dB)	83.9592	83.9413	83.8456	83.5451	65.2555	65.5740	64.7333	63.2888

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